

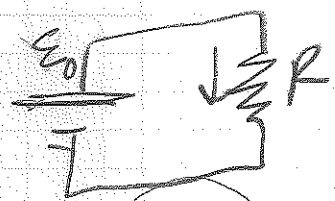
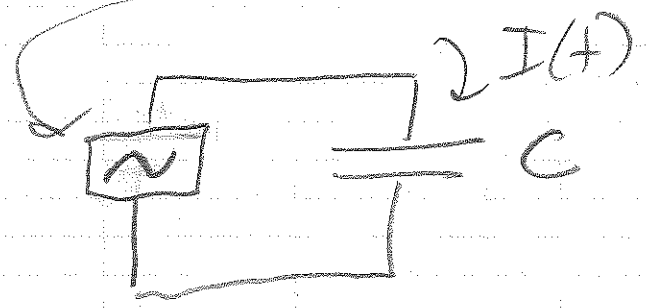
Chapter 8

60 Hz $\omega = 377$ rad/s

DC: $\frac{\epsilon_0}{R}$

AC: $\epsilon_0 \cos(\omega t) = \text{Re}(\epsilon_0 e^{i\omega t})$

$I = \frac{V}{R}$



more subtle relationship.

$I = \frac{V}{R}$ (circled)

$Q = CV$

$IV = \frac{V^2}{R} = I^2 R$

$\int I dt = \frac{I_0}{i\omega} e^{i\omega t} = C \epsilon_0 e^{i\omega t}$

$\text{Re}(I_0 e^{i\omega t})$

$I_0 = i\omega C \epsilon_0$

power always dissipated

same as driver

like $\frac{1}{R}$ (circled)

(DC)

"Admittance"

$\epsilon_0 = \frac{I_0}{i\omega C} = \frac{-i}{\omega C} I_0$

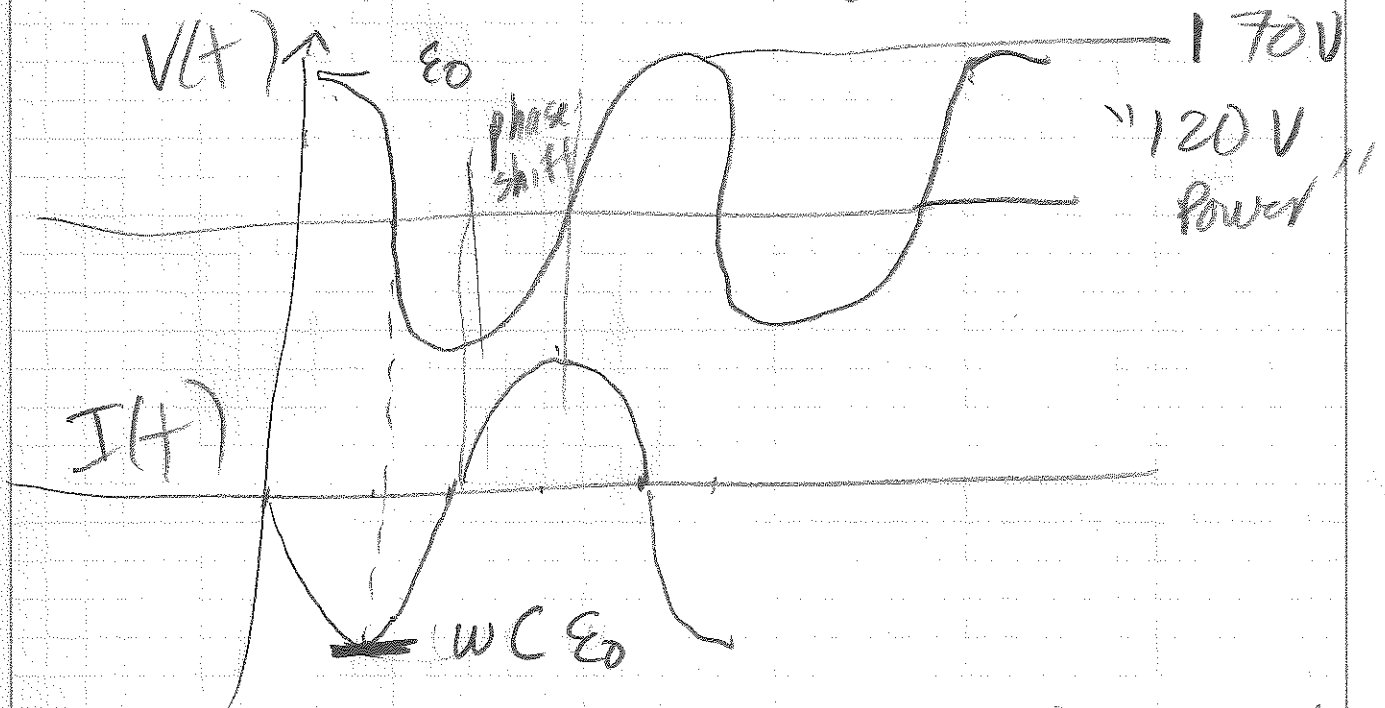
"Impedance"

What does the i mean in I_0 ?

means : $I(t) = \text{Re} (i\omega C \epsilon_0 e^{i\omega t})$

$$e^{i\omega t} = \cos(\omega t) + i\sin(\omega t) = \text{Re} (i\omega C \epsilon_0 \cos(\omega t) - \omega C \epsilon_0 \sin(\omega t))$$

$$I(t) = -\omega C \epsilon_0 \sin(\omega t)$$

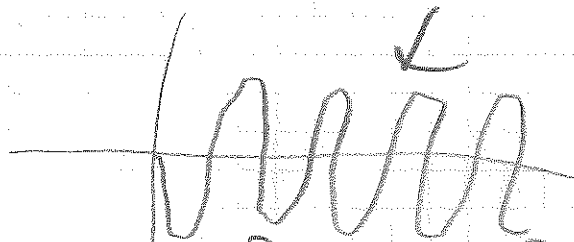


Note $\underline{I(t)V(t)} = -\omega C \epsilon_0 \sin(\omega t) \cdot \epsilon_0 \cos(\omega t)$
 $= \underline{-\frac{\omega C \epsilon_0^2}{2} \sin(2\omega t)}$

instant. power
dissipated



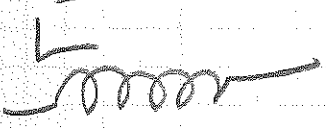
averages to

$\underline{0!}$

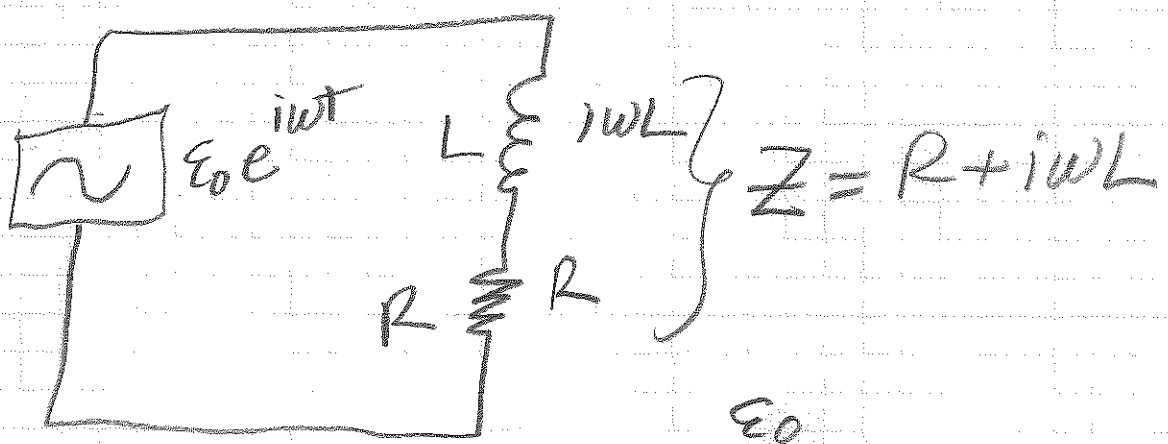


$$\langle V^2 \rangle = \frac{1}{2} \epsilon_0^2 = (120\text{V})^2, \quad \epsilon_0 = \sqrt{2} \times 120 = 170\text{V}$$

Inductor... too

Symbol	Admittance	Impedance
	$1/R$	R
	$i\omega C$	$\frac{-i}{\omega C}$
	$\frac{-i}{\omega L}$	$i\omega L$

Fun ...



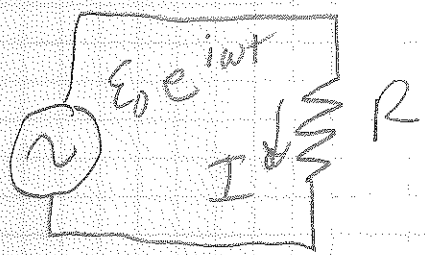
$$I_0 = \frac{E_0}{R + i\omega L} \quad \text{not Lossy}$$

$$= \frac{E_0 (R - i\omega L)}{R^2 + \omega^2 L^2} \quad \text{Lossy}$$

$$\langle I V \rangle = \frac{E_0^2 R}{R^2 + \omega^2 L^2} \times \frac{1}{2} = \frac{V_{rms}^2}{R \left(1 + \omega^2 \frac{L^2}{R^2} \right)}$$

Single Resistive Element

AC Voltage



$$I = \frac{1}{R} V = \frac{1}{R} \epsilon_0 e^{-i\omega t}$$

Key: zoom in on the Power dissipation

$$I(t) V(t) = \frac{1}{R} \epsilon_0 \cos(\omega t) \epsilon_0 \cos(\omega t)$$

$$= \frac{\epsilon_0^2 \cos^2(\omega t)}{R}$$

uh-oh, not
LINEAR!
go back to
REAL PARTS

now, time average!

$$\bar{P} = \frac{\epsilon_0^2}{R} \times \frac{1}{2} \equiv V_{rms}$$

square root

$$V_{rms} = \frac{1}{\sqrt{2}} \epsilon_0 = 120V$$

$$\epsilon_0 = \sqrt{2} \cdot 120V \text{ U.S.} = 169.7V$$

"AC"

$$I_{rms} = \frac{V_{rms}}{R}$$

$$I_{rms} V_{rms} = \frac{V_{rms}^2}{R} = \bar{P}$$

THIS IS TRUE IN GENERAL ONLY FOR RESISTIVE ELEMENTS

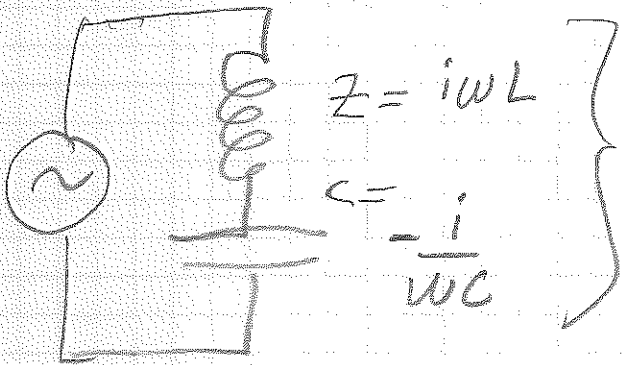
Generally when $I = I_0 e^{i\omega t + \phi}$

$$\bar{P} = I_{rms} V_{rms} \cos \phi$$

Purely imaginary impedances give
 $\phi = \pm \pi/2$
 $+\pi/2$: Capacitor
 $-\pi/2$: Inductor

and $\bar{P} = 0$.

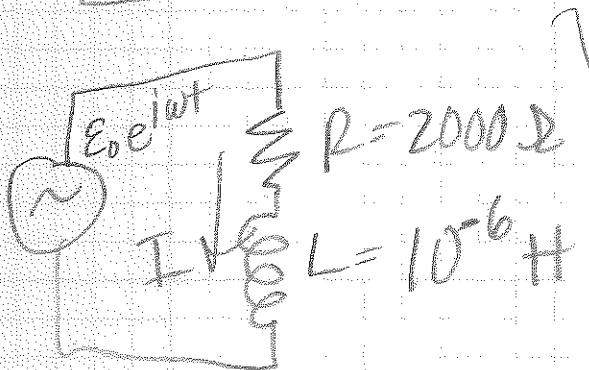
How about



$$Z_{tot} = i\omega L - \frac{i}{\omega C}$$

= purely imaginary.
 NO Power Dissipated.

8.2 (almost)



$$Z = R + i\omega L$$

$$I = \frac{\epsilon_0 e^{i\omega t}}{R + i\omega L}$$

↑
get out of denominator

$$= \frac{\epsilon_0 (R - i\omega L) e^{i\omega t}}{R^2 + \omega^2 L^2}$$

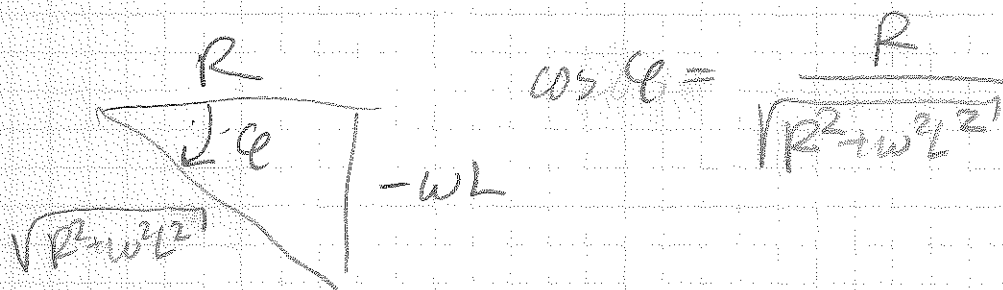
go polar: $R - j\omega L = \sqrt{R^2 + \omega^2 L^2} e^{i \tan^{-1} \left(\frac{-\omega L}{R} \right)}$

$$I = \frac{\epsilon_0}{\sqrt{R^2 + \omega^2 L^2}} e^{i\omega t + i \tan^{-1} \left(\frac{-\omega L}{R} \right)}$$

(take real part)

$$I_{rms} = \frac{1}{\sqrt{2}} \frac{\epsilon_0}{\sqrt{R^2 + \omega^2 L^2}} \quad V_{rms} = \frac{1}{\sqrt{2}} \epsilon_0$$

$$\cos \left(\tan^{-1} \left(\frac{-\omega L}{R} \right) \right)$$



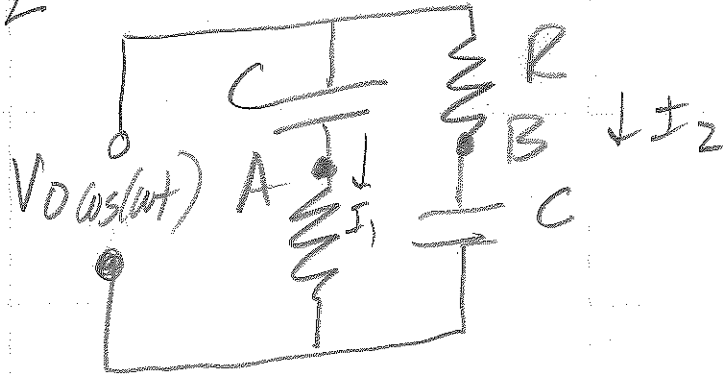
$$\overline{P} = I_{rms} V_{rms} \cos \phi$$

$$= \frac{1}{2} \frac{\epsilon_0^2}{\sqrt{R^2 + \omega^2 L^2}} \frac{R}{\sqrt{R^2 + \omega^2 L^2}}$$

↑
 $R \rightarrow 0$, what
 NO POWER DISSIPATED

THE WEIRDNESS ... $(V_{rms})_R + (V_{rms})_L$
 $\neq V_{rms}$

8.12



$V_{AB} = V_B - V_A$ show $|V_{AB}|^2 = V_0^2$
 for any ω

$$Z_1 = \frac{-j}{\omega C} + R$$

$$Z_2 = R - \frac{j}{\omega C} = Z$$

$I_1 = \frac{V_0 e^{i\omega t}}{\frac{-j}{\omega C} + R} \quad \Rightarrow \quad I_2 = \frac{V_0 e^{i\omega t}}{R - \frac{j}{\omega C}}$
 same!

$$\begin{aligned}
 V_A &= I_1 R \\
 &= \frac{R V_0 e^{i\omega t}}{\frac{-j}{\omega C} + R}
 \end{aligned}$$

$$\begin{aligned}
 V_B &= I_2 \left(\frac{-j}{\omega C} \right) \\
 V_B &= \frac{\frac{-j}{\omega C} V_0 e^{i\omega t}}{R - \frac{j}{\omega C}}
 \end{aligned}$$

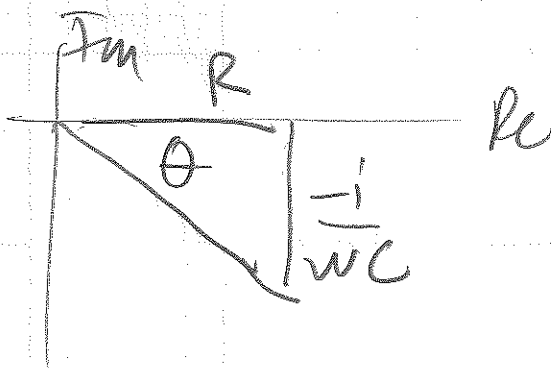
$$V_B - V_A = \frac{\left(\frac{-j}{\omega C} - R \right)}{\left(R - \frac{j}{\omega C} \right)} V_0 e^{i\omega t}$$

$$= -\frac{\left(R + \frac{j}{\omega C}\right)}{\left(R - \frac{j}{\omega C}\right)} V_0 e^{j\omega t}$$

$$= -\frac{z^*}{z} V_0 e^{j\omega t}$$

"a pure phase" $\rightarrow \left| \frac{z^*}{z} \right| = \frac{\sqrt{R^2 + \left(\frac{1}{\omega C}\right)^2}}{\sqrt{R^2 + \left(\frac{1}{\omega C}\right)^2}} = 1$

$$\frac{z^*}{z} = \frac{-e^{-j\theta}}{e^{j\theta}} = -e^{-2j\theta}$$



$$\theta = \tan^{-1}\left(\frac{-1}{\omega CR}\right)$$

Really, $\text{Re}(V_B - V_A)$
 $= \text{Re}\left[-V_0 e^{j(\omega t - 2\theta)}\right]$

$$\text{Re}(V_B - V_A) = -V_0 \cos\left(\omega t - 2 \tan^{-1}\left(\frac{1}{\omega CR}\right)\right)$$

$$2 \tan^{-1} \left(\frac{1}{wCR} \right) = \frac{\pi}{2} = 90^\circ$$

$$\tan^{-1} \left(\frac{1}{wCR} \right) = \frac{\pi}{4} = 45^\circ$$

$$wCR = 1$$

$$w = \frac{1}{RC}$$

Electromagnetic Equations

$$\text{div } \vec{E} = 4\pi \rho$$

$$\text{div } \vec{E} = \frac{\partial E_x}{\partial x} + \frac{\partial E_y}{\partial y} + \frac{\partial E_z}{\partial z}$$

$$\rho = \frac{\text{local charge}}{\text{volume}}$$

$$\text{div } \vec{J} = -\frac{\partial \rho}{\partial t}$$

$$\vec{J} = \frac{\text{charge}}{\text{area} \times \text{time}} \times (\text{direction})$$

always true.

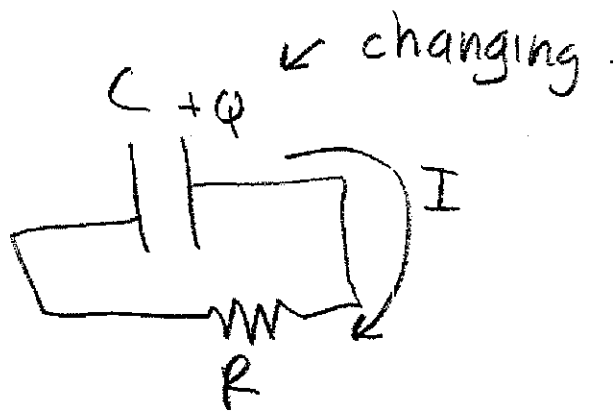
when \vec{J} is constant in time ($\frac{\partial \vec{J}}{\partial t} = 0$)

$$\text{curl } \vec{B} = \frac{4\pi}{c} \vec{J}$$

$$\text{curl } \vec{B} = \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ B_x & B_y & B_z \end{vmatrix}$$

Contradiction

suppose $\frac{\partial \rho}{\partial t} \neq 0$

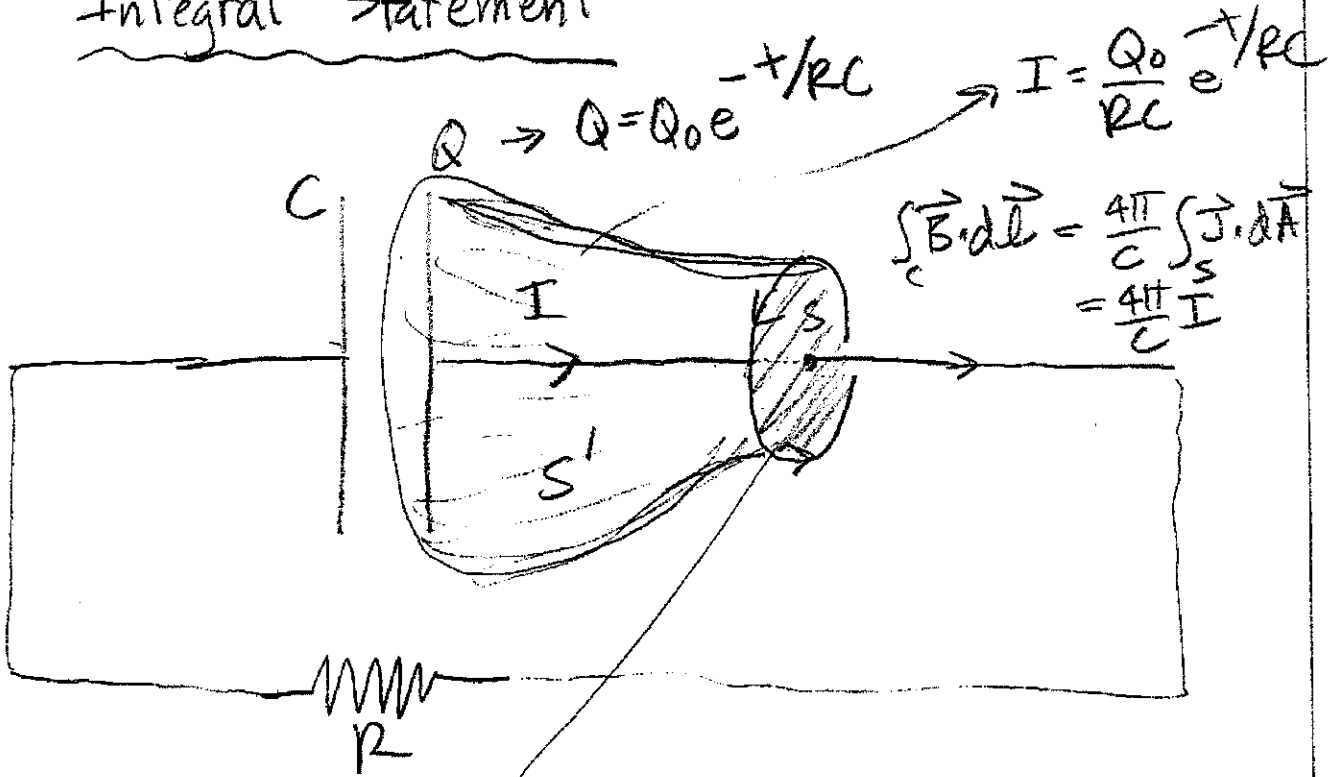


then $\text{div } \vec{J} = -\frac{\partial \rho}{\partial t} \neq 0$

but... $\text{div}(\text{curl } \vec{B}) = 0 = \frac{4\pi}{c} \text{div } \vec{J}$
 (no sources)

That's the contradiction.

Integral Statement



but that is not the only surface one could use!

$$\int_{S'} \vec{J} \cdot d\vec{A} = 0 !$$

so,

$$\text{curl } \vec{B} = \frac{4\pi}{c} \vec{J} + \text{(new term)}$$

↑
some $\frac{\partial(\quad)}{\partial t}$

Look at $\text{curl } \vec{E} = -\frac{1}{c} \frac{\partial \vec{B}}{\partial t}$

recall symmetry of \vec{E}, \vec{B} in "Lorentz" transformation

$$\vec{E}'_{\parallel} = \vec{E}_{\parallel}$$

$$\vec{B}'_{\parallel} = \vec{B}_{\parallel}$$

$$\vec{E}'_{\perp} = \gamma(\vec{E}_{\perp} + \vec{\beta} \times \vec{B}_{\perp})$$

$$\vec{B}'_{\perp} = \gamma(\vec{B}_{\perp} - \vec{\beta} \times \vec{E}_{\perp})$$

↗ - sign.
↖

$$\text{curl } \vec{B} = \frac{4\pi}{c} \vec{J} + \frac{1}{c} \frac{\partial \vec{E}}{\partial t}$$

changing \vec{E} field,
integrated across surface,
gives a "circular" \vec{B}
(like E.M.F.)
"LENZ" \Rightarrow other way.

$$\text{curl } \vec{E} = -\frac{1}{c} \frac{\partial \vec{B}}{\partial t}$$

$$\text{curl } \vec{B} = \frac{1}{c} \frac{\partial \vec{E}}{\partial t} + \frac{4\pi}{c} \vec{J}$$

$$\text{div } \vec{E} = 4\pi \rho$$

$$\text{div } \vec{B} = 0$$

Maxwell's Equations (CGS)

MKS/SI:

$$\text{curl } \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

$$\text{curl } \vec{B} = \mu_0 \epsilon_0 \frac{\partial \vec{E}}{\partial t} + \mu_0 \vec{J}$$

$$\text{div } \vec{E} = \epsilon_0 \rho$$

$$\text{div } \vec{B} = 0$$

$$\mu_0 \epsilon_0 = \frac{1}{c^2}$$